

Benefits of coherent demodulation for eavesdropping on HDMI emissions

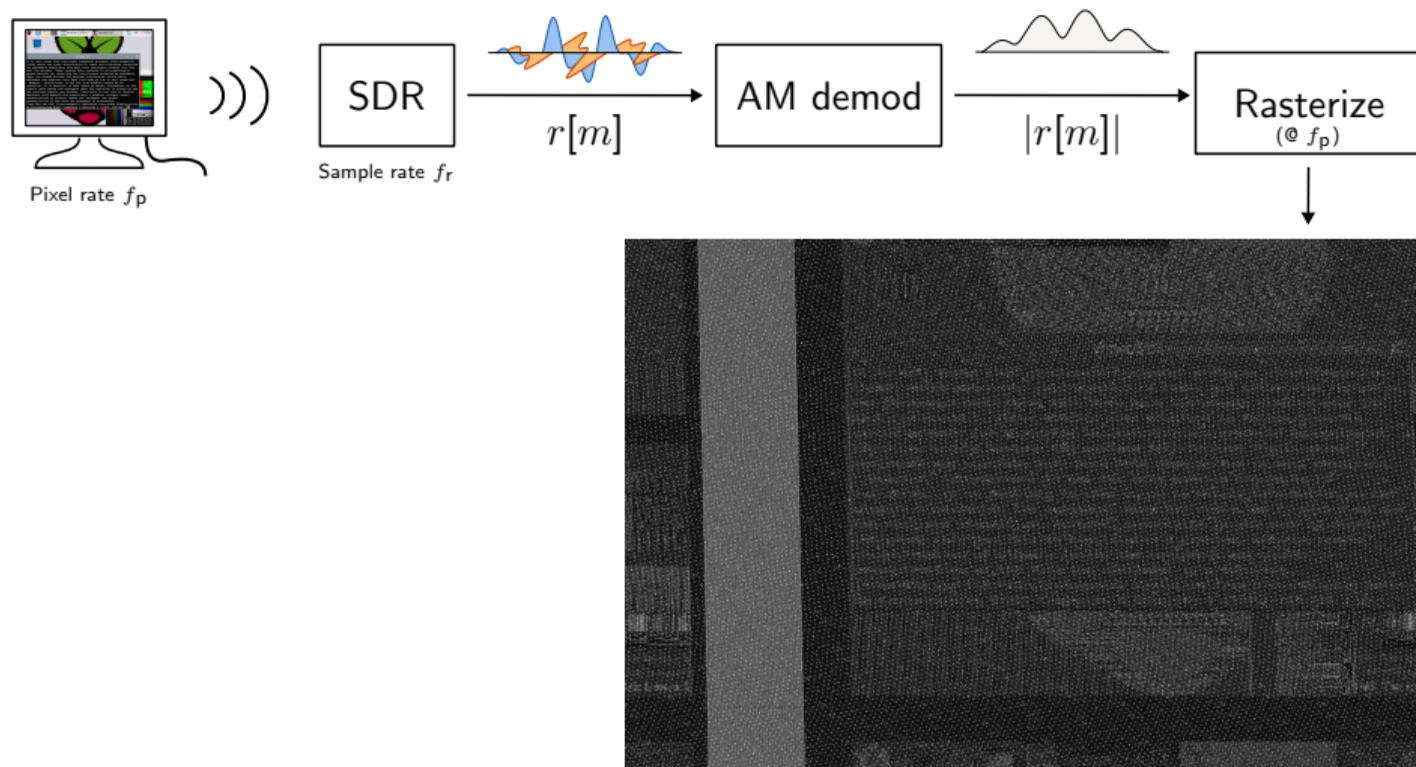
Dimitrije Erdeljan, Markus G. Kuhn

Department of Computer Science and Technology
University of Cambridge

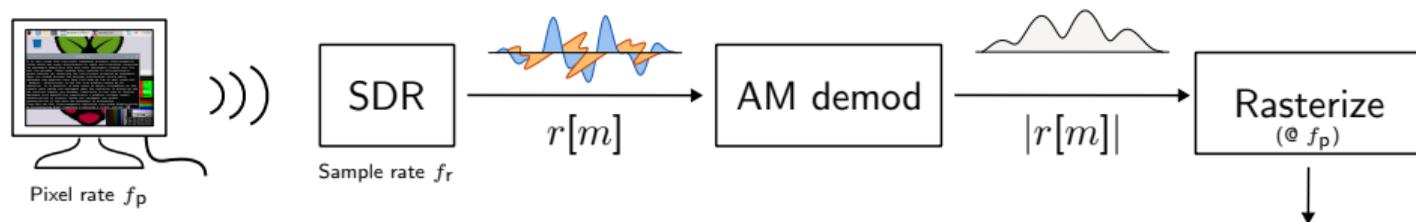


EMC Europe 2024

A typical TEMPEST attack



A typical TEMPEST attack

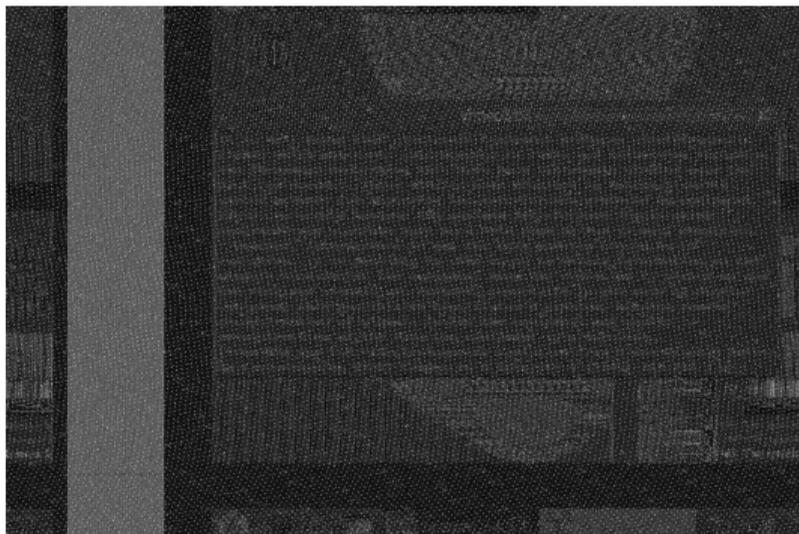


- ▶ Estimate pixel rate f_p , e.g. from the autocorrelation

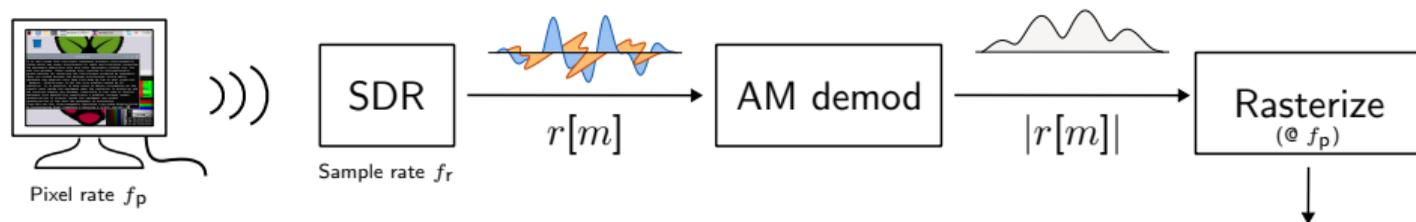
$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*$$

$$\hat{d} = \underset{f_{v,\perp} \leq f_r/d \leq f_{v,\top}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$



A typical TEMPEST attack



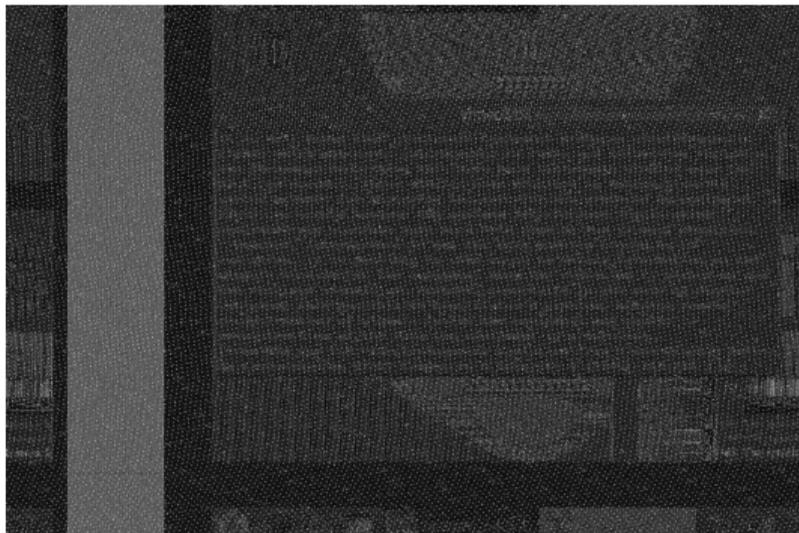
- ▶ Estimate pixel rate f_p , e.g. from the autocorrelation

$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*:$$

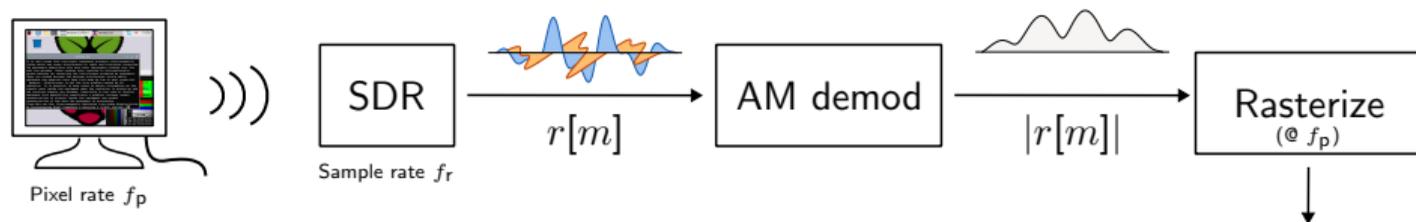
$$\hat{d} = \underset{f_{v,\perp} \leq f_r/d \leq f_{v,\top}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$

- ▶ Resample to $f_s = k \cdot f_p$ for $k \in \mathbb{N}$



A typical TEMPEST attack



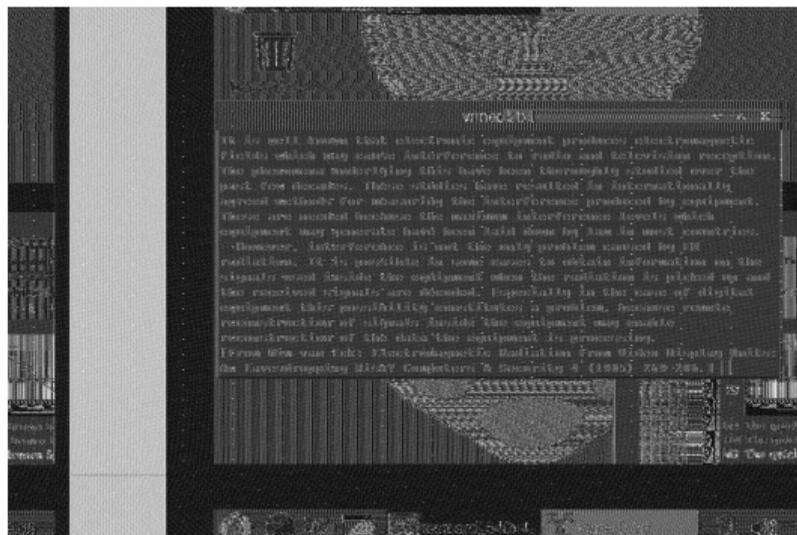
- ▶ Estimate pixel rate f_p , e.g. from the autocorrelation

$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*$$

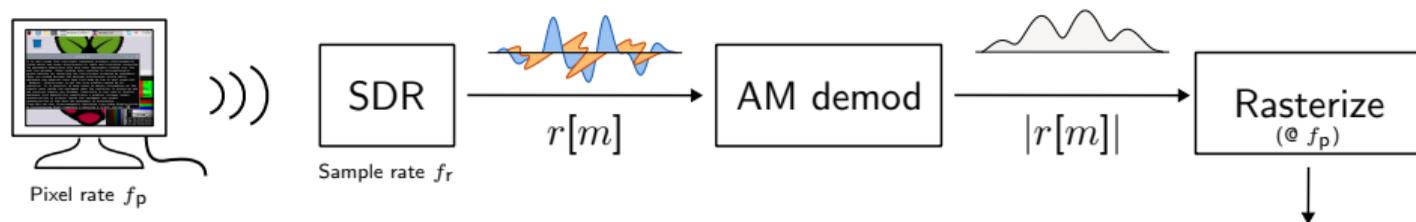
$$\hat{d} = \underset{f_v, \pm \leq f_r/d \leq f_v, \tau}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$

- ▶ Resample to $f_s = k \cdot f_p$ for $k \in \mathbb{N}$
- ▶ Average several rasterized frames



A typical TEMPEST attack



- ▶ Estimate pixel rate f_p , e.g. from the autocorrelation

$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*:$$

$$\hat{d} = \underset{f_{v,\perp} \leq f_r/d \leq f_{v,\top}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

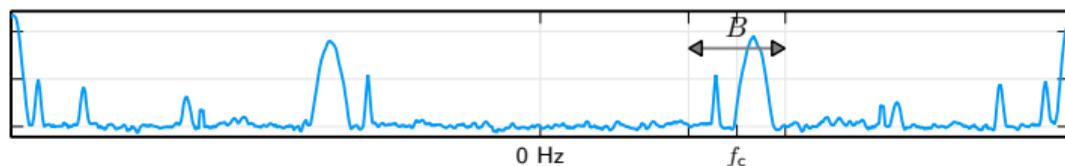
$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$

- ▶ Resample to $f_s = k \cdot f_p$ for $k \in \mathbb{N}$
- ▶ Average several rasterized frames
- ▶ Align image

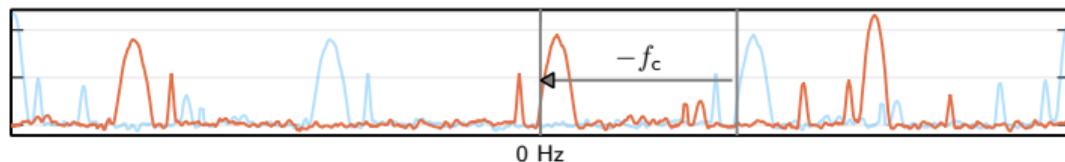


Software-defined radio receiver

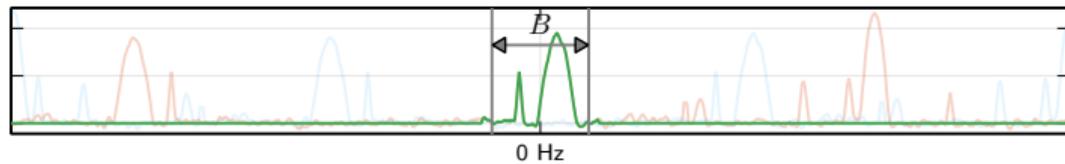
Antenna waveform (shown as Fourier spectrum) $s_0(t)$:



Downconvert: $s_d(t) = s_0(t) \cdot e^{-2\pi j f_c t}$



Lowpass filter: $s_f(t) = \int s_d(t - \tau)g(\tau)d\tau$



Finally, output sampled $r[m] = s_f(m/f_r)$.

Rasterizing complex-valued signals: amplitude demodulation

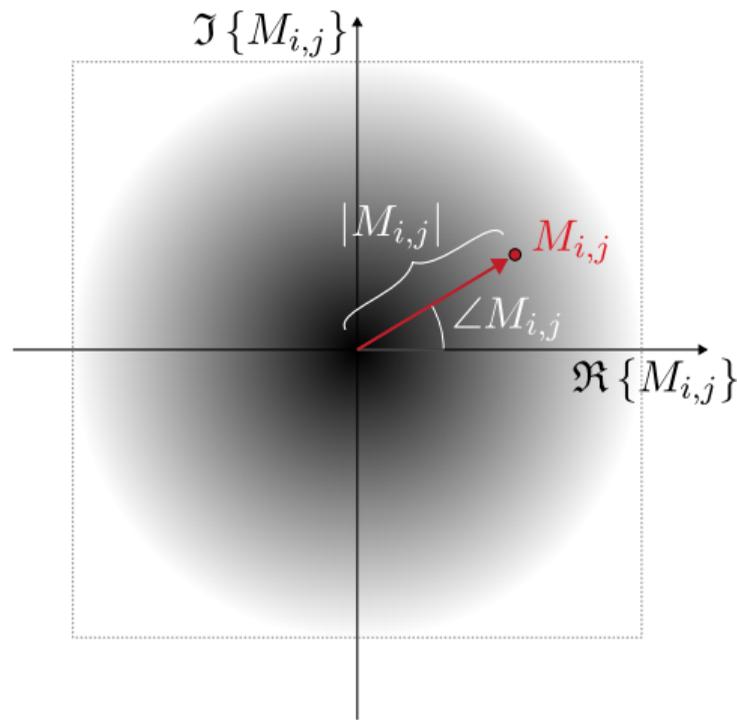
Most eavesdropping demonstrations amplitude demodulate samples $M_{i,j} \in \mathbb{C}$ and visualise them as grayscale pixels.

For example, mapping 1% and 99% quintiles to black and white:

$$\text{Gray} \left(\frac{|M_{i,j}| - q_{1\%}}{q_{99\%} - q_{1\%}} \right)$$

This discards phase information $\angle M_{i,j}$.

The quick brown fox

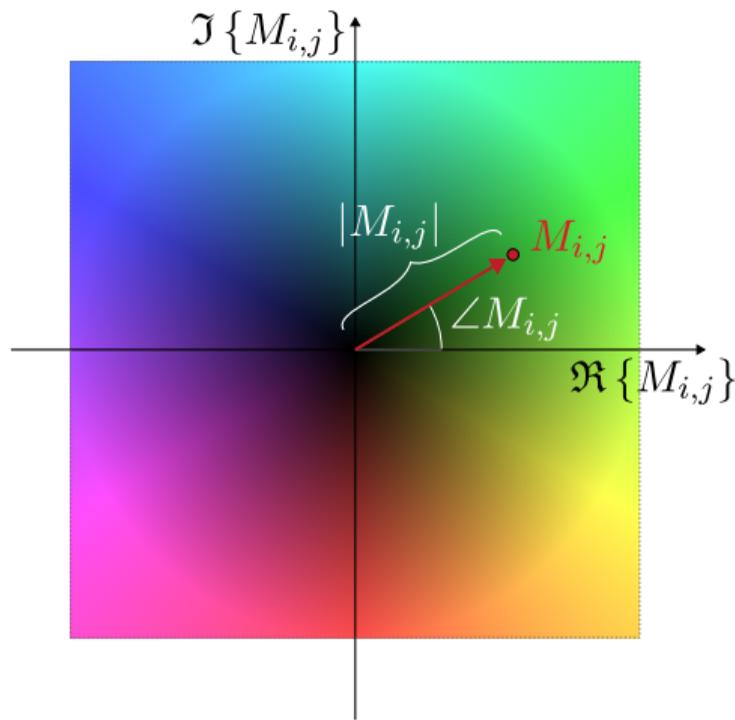


Rasterizing complex-valued signals: HSV visualisation

Using the HSV (hue, saturation, value) colour space allows us to also show the phase:

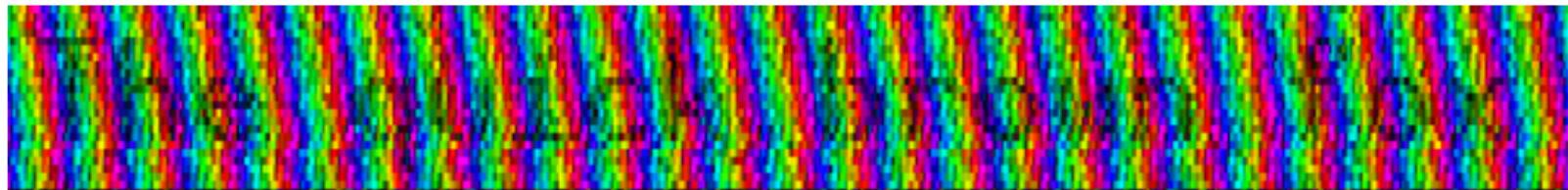
$$\text{HSV} \left(\angle M_{i,j}, S, \frac{|M_{i,j}| - q_{1\%}}{q_{99\%} - q_{1\%}} \right)$$

(We leave the saturation coordinate S as a user preference.)

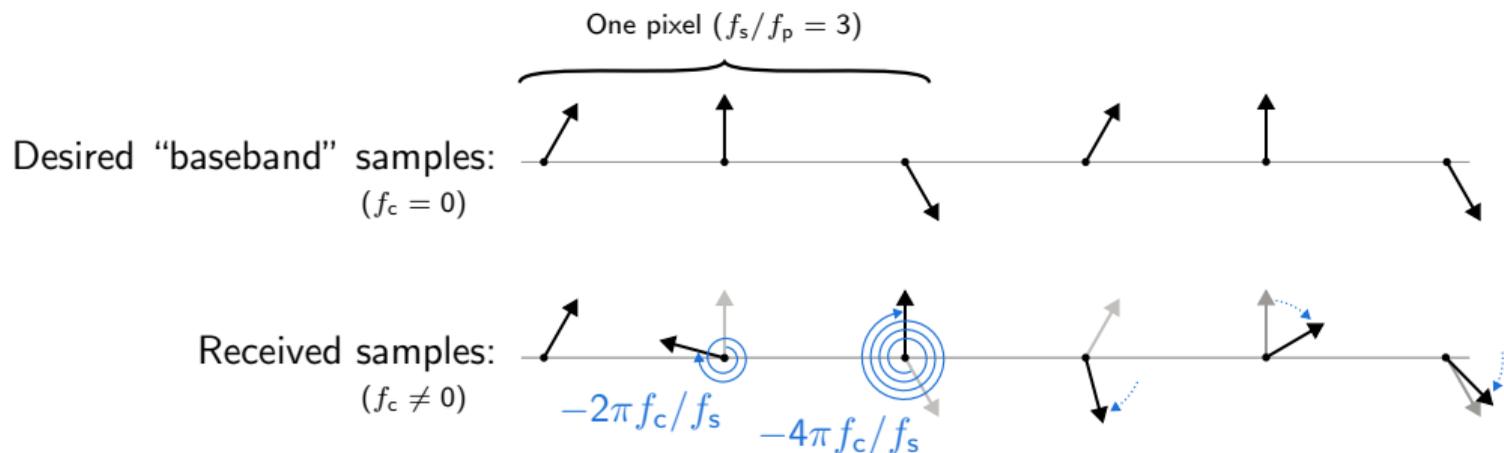


First rasterization attempt

Directly rasterizing an SDR-received signal produces a “rainbow-banding” image:

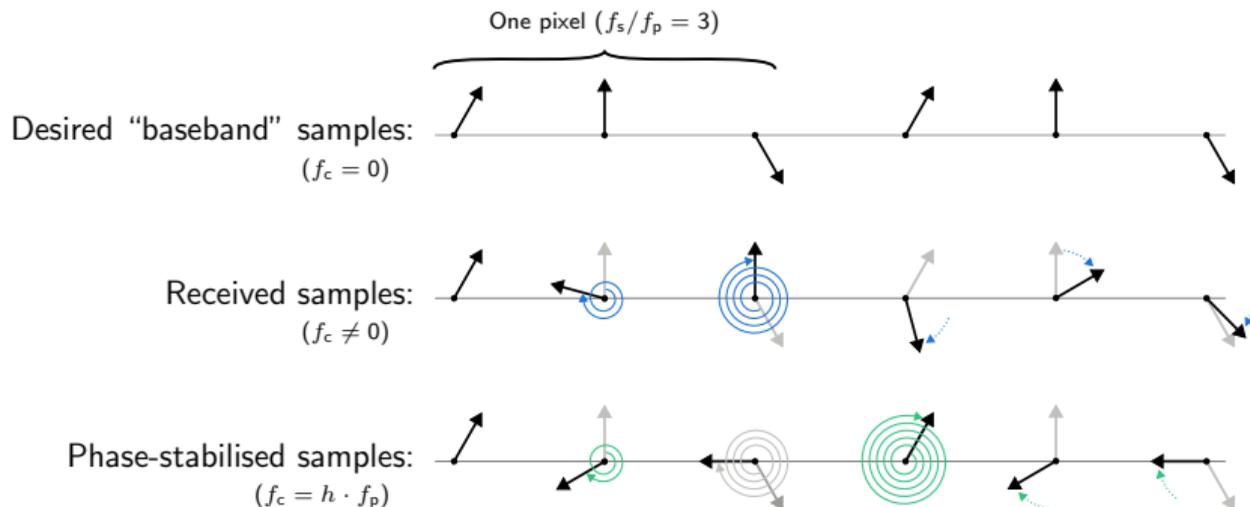


This is due to SDR downconversion from the antenna waveform $s_0(t)$ to $e^{-2\pi j f_c t} \cdot s_0(t)$.



Obtaining consistent phase angles

Shift the centre frequency to a harmonic $h \cdot f_p$ of the pixel frequency:



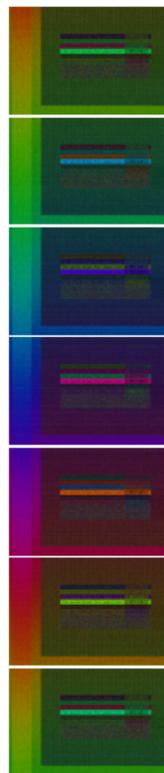
We combine **frequency shifting** $f_c \rightarrow h \cdot f_p$ with **resampling** to $f_s = k \cdot f_p$:

$$s[n] \approx s_f \left(\frac{n + \lambda}{f_s} \right) \cdot e^{2\pi j(f_c - h f_p)n / f_s}$$

Obtaining consistent phase angles

Some drift still remains over longer intervals.

Coherent averaging requires consistent phase across many frames, i.e. a more accurate f_p estimate.



Accurate f_p estimation

We improve the f_p estimate several times until convergence, by iterating over three steps:

- 1 Resampling and frequency-shifting $f_c \rightarrow h \cdot f_p$:

$$s[n] \approx s_f \left(\frac{n + \lambda}{f_s} \right) \cdot e^{2\pi j(f_c - h f_p)n / f_s}$$

- 2 Computing the autocorrelation:

$$R_{s,s}[d] = \sum_n s[n] \cdot s[n + d]^*$$

- 3 Updating the f_p estimate, with a fine-tuning term which measures phase drift between frames:

$$f_p := f_p \cdot \left(\frac{k\omega_t h_t}{\hat{d}} + \frac{k\angle R_{s,s}[\hat{d}]}{2\pi h \hat{d}} \right)$$

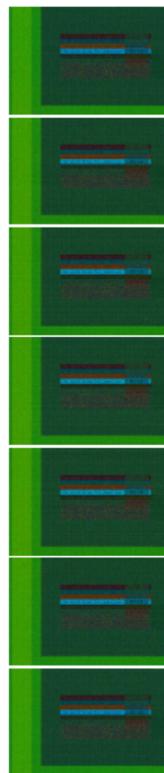
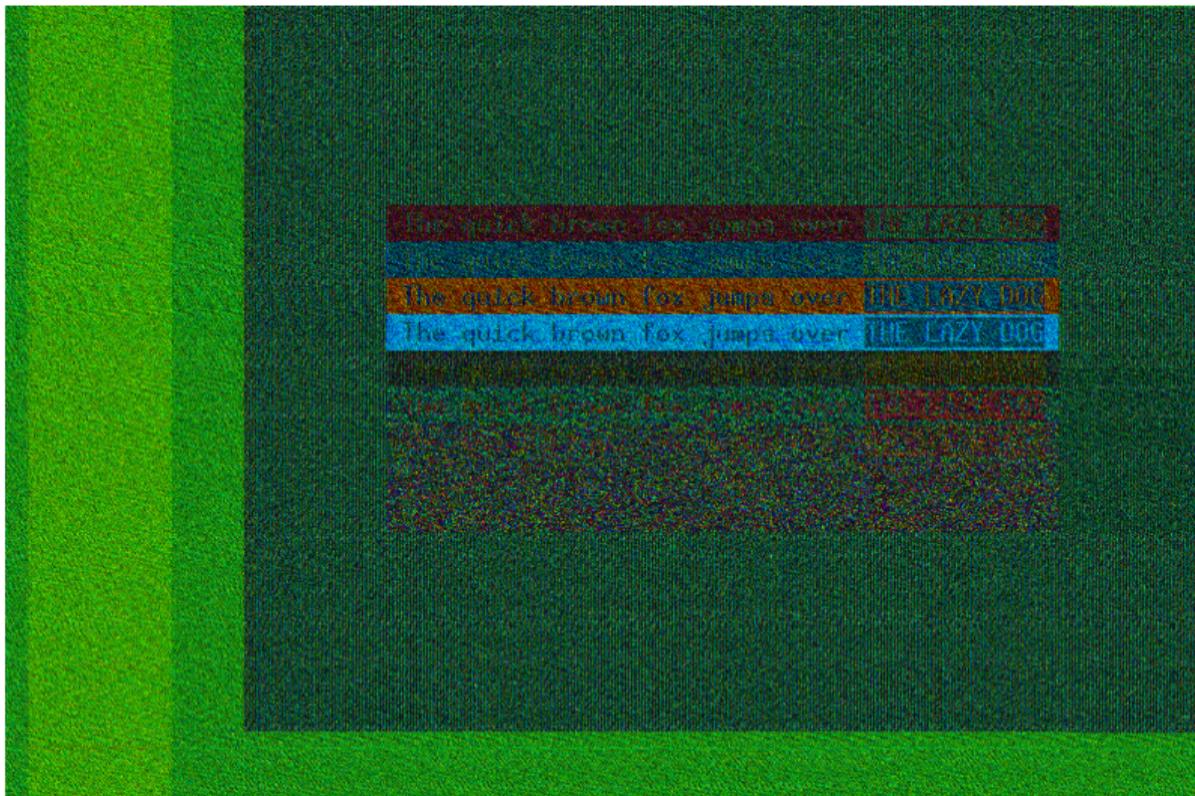
Iterations	f_p
0	25.200000000 MHz
1	25.200096064 MHz
2	25.200096764 MHz
3	25.200096794 MHz
4	25.200096793 MHz
5	25.200096788 MHz
6	25.200096788 MHz

$$\hat{d} = \underset{f_{v,\perp} \leq f_r / d \leq f_{v,\top}}{\operatorname{argmax}} |R_{s,s}[d]|^2$$

In later iterations, we can also search for the correlation peak at larger multiples of the frame period.

Obtaining consistent phase angles (with accurate f_p)

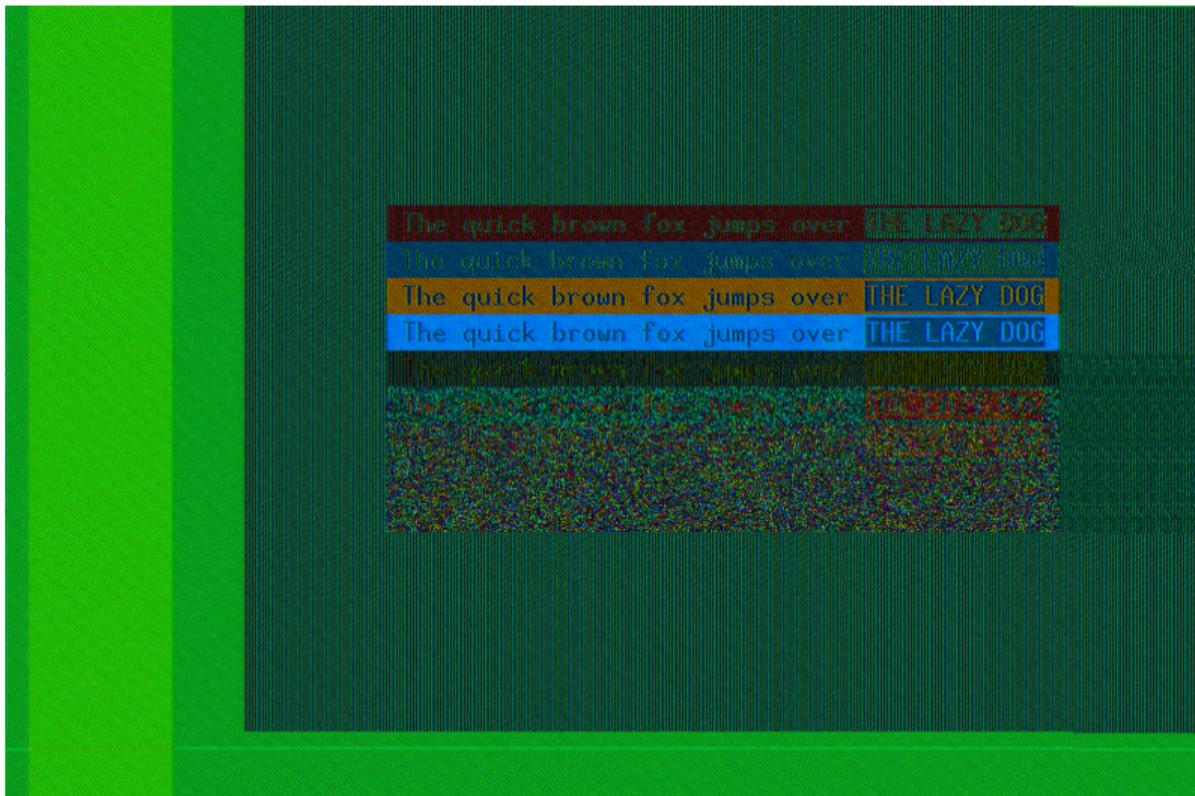
With the more accurate f_p , the phase now stays consistent across frames.



Consistent phase angles enable coherent averaging

We can now periodically average unrotated $M_{i,j} \in \mathbb{C}$ to reduce noise.

This image was rasterized from 30 averaged frames (≈ 0.5 s long).



Does the phase provide new information?

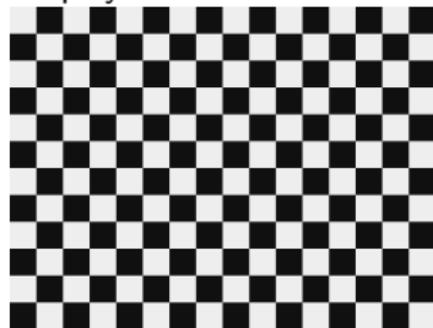
To demonstrate how phase information can help with distinguishing colours, we used a test image with two grayscale colours: #101010 and #eeeeee.

These are TMDS-encoded in HDMI as complementary bit sequences:

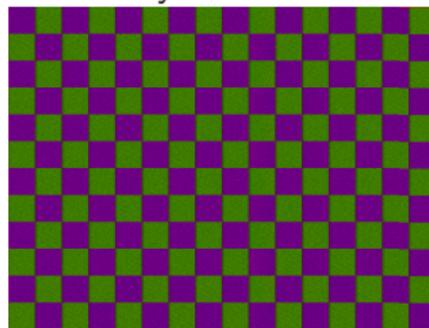
- ▶ #101010 → 0111110000
- ▶ #eeeeee → 1000001111

The resulting emissions therefore differ only in their sign.

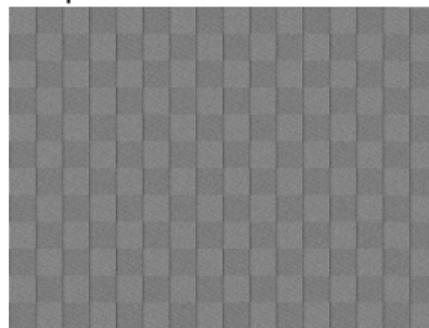
Displayed:



Coherently demodulated:



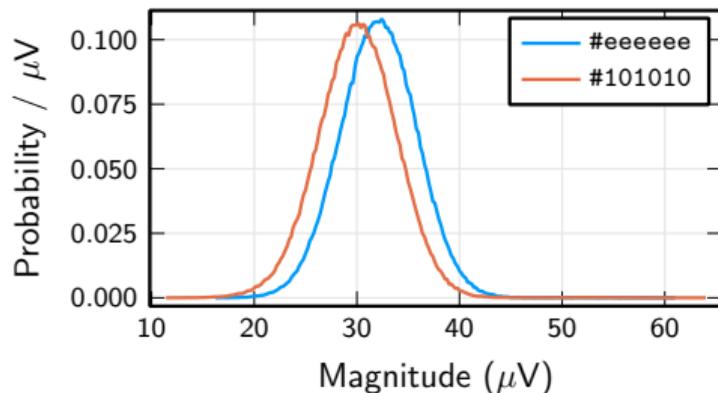
Amplitude demodulated:



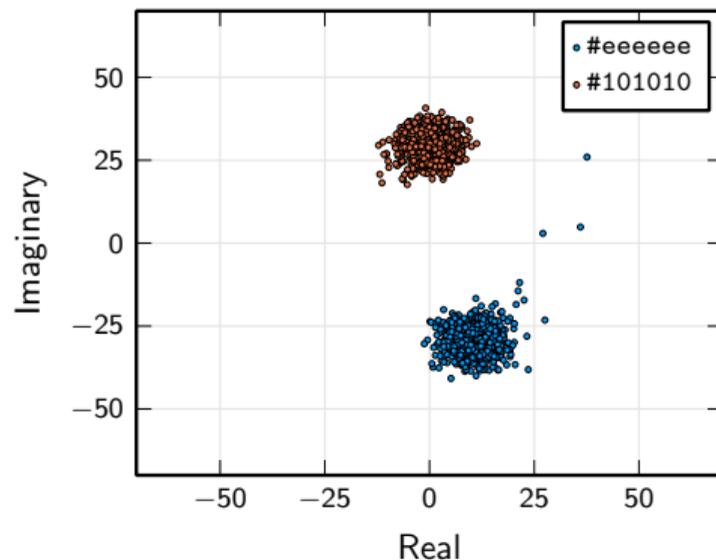
Does the phase provide new information?

The distributions of amplitudes $|M_{i,j}|$ for these two colours overlap significantly, while those for the full $M_{i,j} \in \mathbb{C}$ do not:

Amplitude distribution ($|M_{i,j}|$):



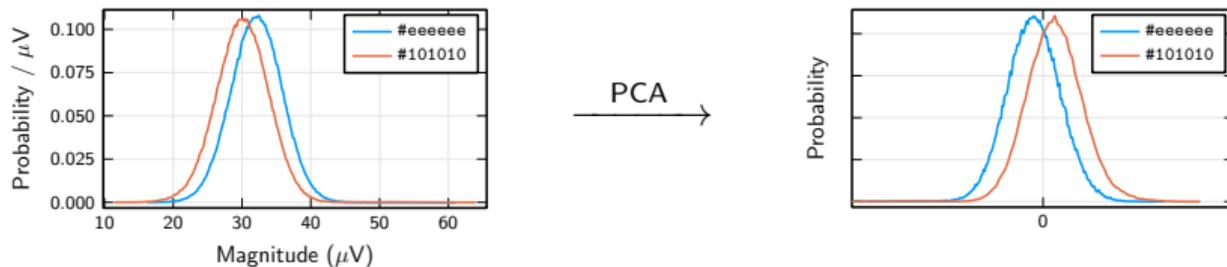
Complex distribution ($M_{i,j}$):



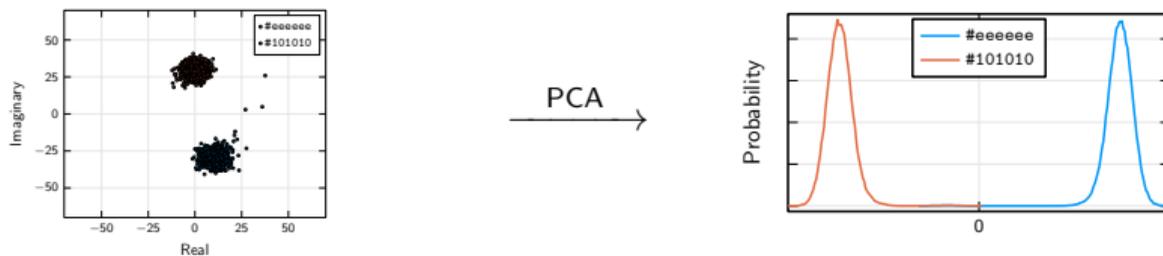
Comparing distributions using dimensionality reduction

We can better distinguish colours by combining information from all $k = 3$ samples for a pixel with dimensionality reduction, e.g. using Principal Component Analysis (PCA).

Amplitude demodulation (3-dimensional PCA on $|M_{i,3j}|, |M_{i,3j+1}|, |M_{i,3j+2}|$):



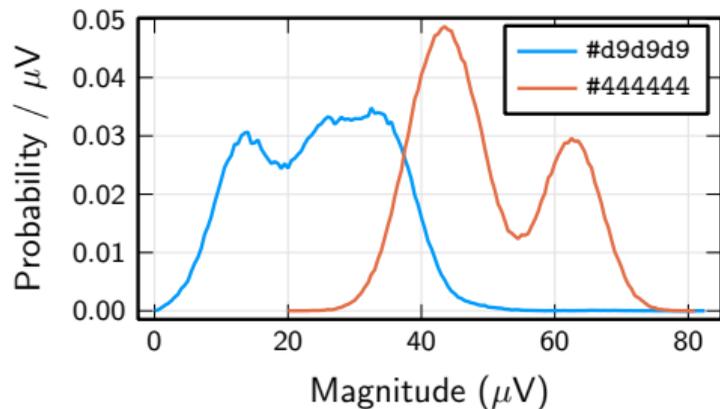
Coherent demodulation (6-dimensional PCA on $\Re\{M_{i,3j}\}, \Im\{M_{i,3j}\}, \dots, \Im\{M_{i,3j+2}\}$):



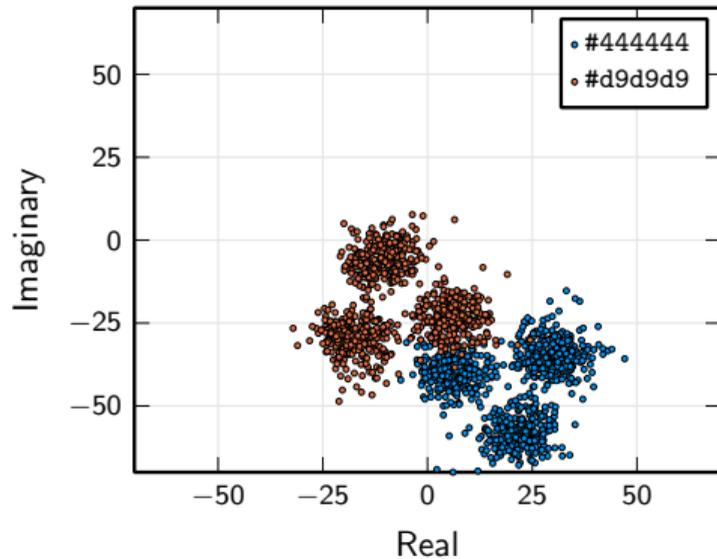
A more complex encoding

For another randomly chosen pair of colours, their more varied TMDS encodings result in different distributions for each sample position in a pixel:

Amplitude distribution ($|M_{i,j}|$):



Complex distribution ($M_{i,j}$):



PCA-based rasterization

Rasterizing the projection on the largest-eigenvalue PCA vector can produce grayscale images with better contrast than amplitude demodulation.

1 Black and white:

The quick brown fox jumps over THE LAZY DOG



2 Maximum bit transition contrast:

The quick brown fox jumps over THE LAZY DOG



3 Complementary TMD5 encoding:

The quick brown fox jumps over THE LAZY DOG



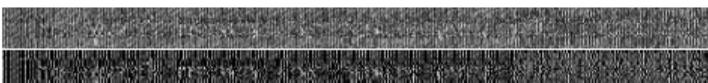
4 Maximum contrast if AM demodulated:

The quick brown fox jumps over THE LAZY DOG



5 Minimum contrast if AM demodulated:

The quick brown fox jumps over THE LAZY DOG



6 Low nibble random:

The quick brown fox jumps over THE LAZY DOG



(Top to bottom: displayed, amplitude demodulated, PCA)

- ▶ Phase information can be included in the rasterized image as hue in the HSV colour space
- ▶ Shifting the centre frequency precisely to a harmonic of the pixel clock stabilises phase and allows periodic averaging in the complex domain
- ▶ We can precisely estimate the target's pixel-clock frequency using both position and phase of the peak of the complex-valued autocorrelation sequence
- ▶ Preserving phase information helps better discriminate between colours
- ▶ With multiple samples per pixel, dimensionality reduction can further improve contrast